

Electrical Circuits (2)



Lecture 2 Resonance

Dr.Eng. Basem ElHalawany

Main Topics

1. Resonance
2. Magnetically Coupled Circuits
3. Three-Phase Circuits
4. Transient Analysis



References

- A. Fundamentals of Electric Circuits (Alexander and Sadiku)**
- B. Principles of Electric Circuits (Floyd)**
- C. Circuit Analysis – Theories and Practice (Robinson & Miller)**
- D. Introductory Circuit Analysis (Boylestad)**



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Resonance

Circuits with both inductance and capacitance can exhibit a property called “Resonance” which is important in many applications

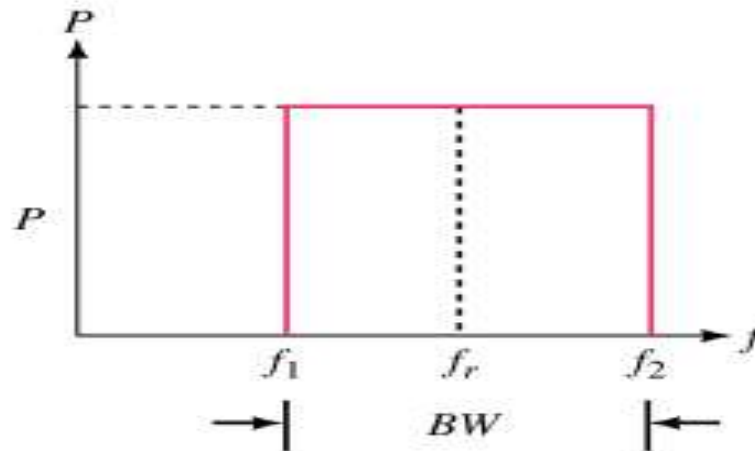
- Resonance is the basis for frequency selectivity in communication systems
- The ability of a radio or TV receiver to select a certain frequency (station) and at the same time eliminate frequencies from other stations is based on the principle of resonance

**In this chapters
we will observe how resonant circuits are able to pass a desired
range of frequencies from a signal source to a load.**



Resonance

- ✓ In order to obtain all the transmitted energy for a given radio station or television channel, we would like a circuit to have the frequency response shown in Figure 21-1.a:



(a) Ideal frequency response curve

- f_r : center frequency = station carrier frequency
- BW : bandwidth of the station = The difference between the upper and lower frequencies that we would like to pass

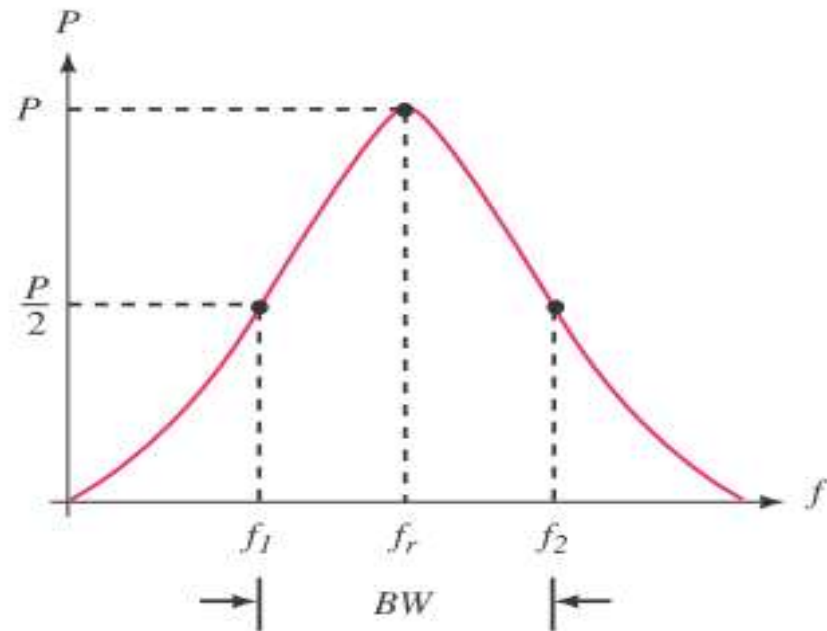
A circuit having an ideal frequency response would pass all frequency components in a band between f_1 and f_2 , while rejecting all other frequencies.

Resonance

Whereas there are various configurations of resonant circuits, they all have several common characteristics.

1. The resonant circuit consists of at least an **inductor** and a **capacitor** together with a **voltage or current source**.
2. Have a bell-shaped response curve centered at some resonant frequency as in shown in figure
3. This curve indicates that power will be a maximum at f_r and varying the frequency in either direction results in a reduction of the power.

The bandwidth = the difference between the half-power points on the response curve of the filter.



(b) Actual response curve of a resonant circuit

21.1 Series Resonance

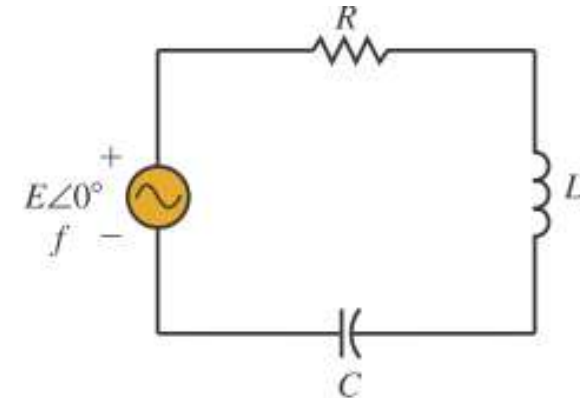
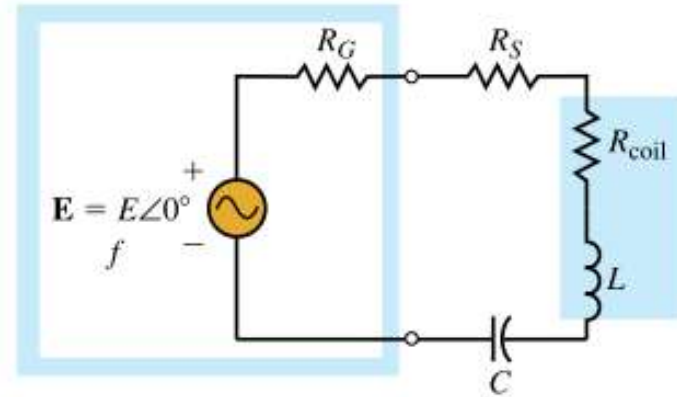
- R_G : Generator resistance
- R_S : Series resistance
- R_{coil} : Inductor coil resistance

In this circuit, the total resistance is expressed as

$$R = R_G + R_S + R_{coil}$$

The total impedance is given by:

$$\begin{aligned} Z_T &= R + jX_L - jX_C \\ &= R + j(X_L - X_C) \\ &= R + j\left(\omega L - \frac{1}{\omega C}\right) \end{aligned}$$



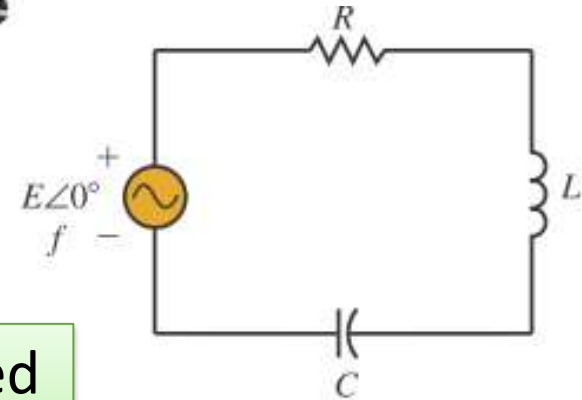
Resonance is a condition in an RLC circuit in which the capacitive and inductive reactances are equal in magnitude, thereby resulting in a purely resistive impedance.

21.1 Series Resonance

By setting the reactance of the capacitor and inductor equal to one another, the total impedances given by:

$$Z_T = R$$

The value of ω that satisfies this condition is called the resonant frequency



$$\omega L = \frac{1}{\omega C}$$

$$\omega^2 = \frac{1}{LC}$$

$$\omega_s = \frac{1}{\sqrt{LC}} \quad (\text{rad/s})$$

$$f_s = \frac{1}{2\pi\sqrt{LC}} \quad (\text{Hz})$$

OR

$$\omega_0 = \frac{1}{\sqrt{LC}} \text{ rad/s}$$



ANALYSIS OF SERIES RLC CIRCUITS

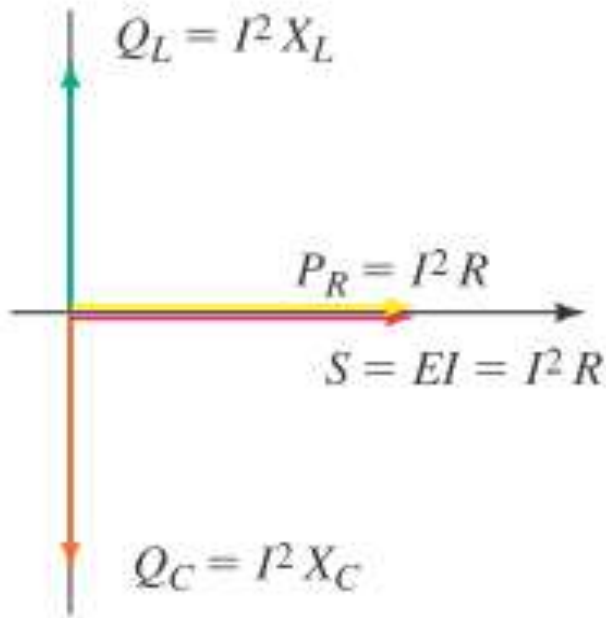
At resonance the total impedances given by:

$$Z_T = R$$

At resonance, the total current in the circuit is determined from Ohm's law as

$$\mathbf{I} = \frac{\mathbf{E}}{Z_T} = \frac{E \angle 0^\circ}{R \angle 0^\circ} = \frac{E}{R} \angle 0^\circ$$

The voltage across each of the elements in the circuit as follows:



$$V_R = IR \angle 0^\circ$$

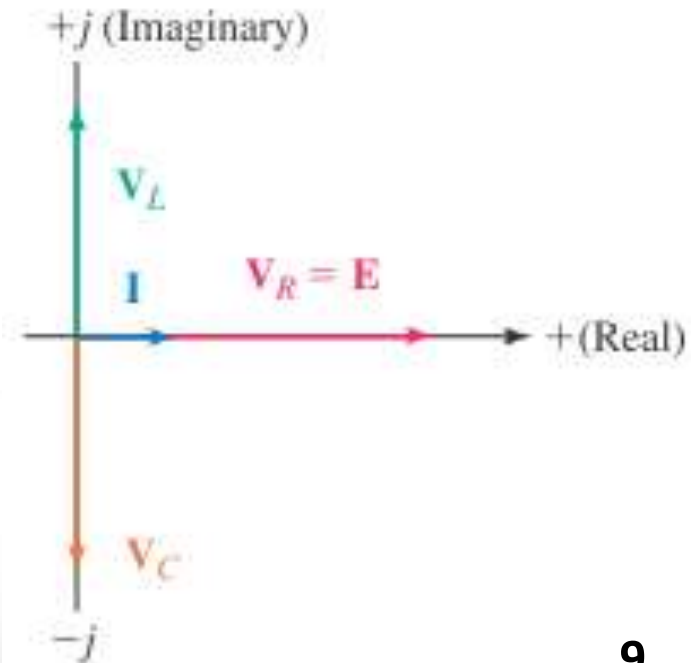
$$V_L = IX_L \angle 90^\circ$$

$$V_C = IX_C \angle -90^\circ$$

$$P_R = I^2 R \quad (\text{W})$$

$$Q_L = I^2 X_L \quad (\text{VAR})$$

$$Q_C = I^2 X_C \quad (\text{VAR})$$



Impedance of a Series Resonant Circuit **versus** Frequency

Because the impedances of (L and C) are dependent upon frequency, the total impedance of a series resonant circuit must similarly vary with frequency

$$\begin{aligned}Z_T &= R + j\omega L - j\frac{1}{\omega C} \\ &= R + j\left(\frac{\omega^2 LC - 1}{\omega C}\right)\end{aligned}$$

Impedances Magnitude:

$$Z_T = \sqrt{R^2 + \left(\frac{\omega^2 LC - 1}{\omega C}\right)^2}$$

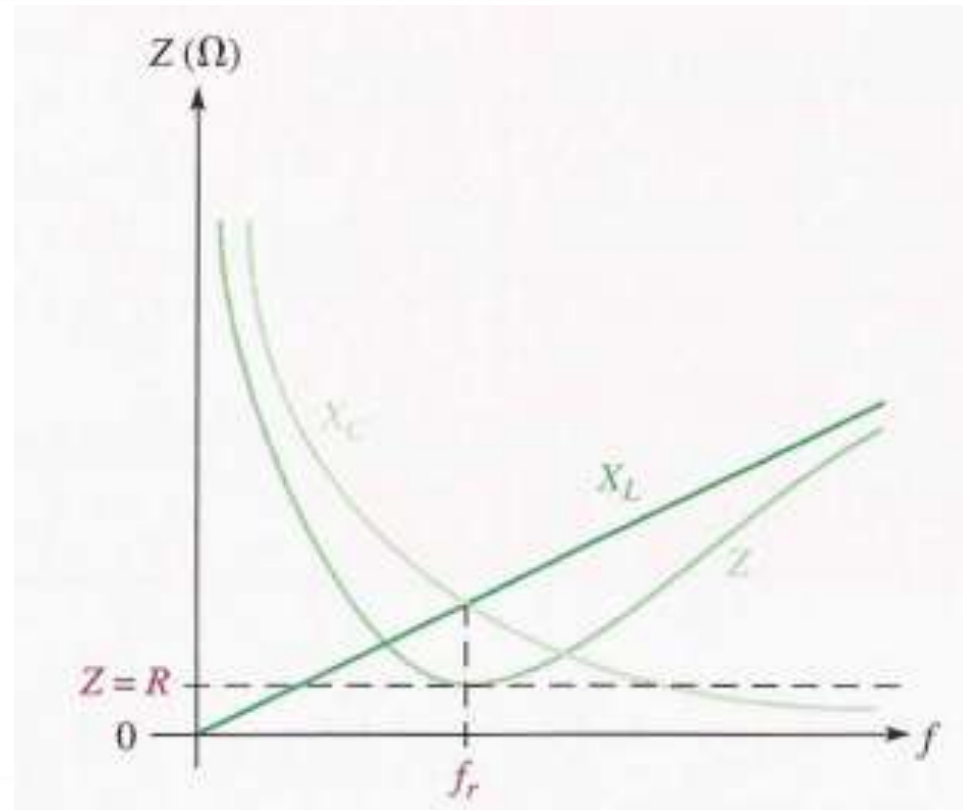
Impedances Angle:

$$\theta = \tan^{-1}\left(\frac{\omega^2 LC - 1}{\omega RC}\right)$$

When $\omega = \omega_S$:

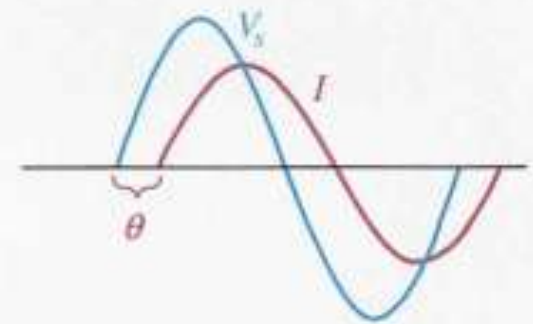
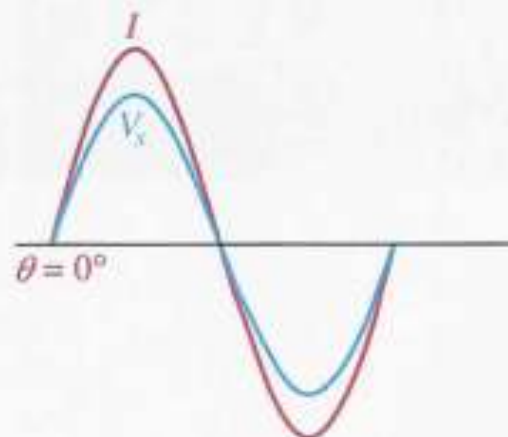
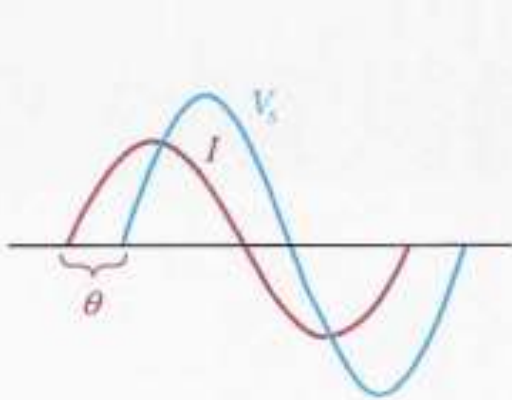
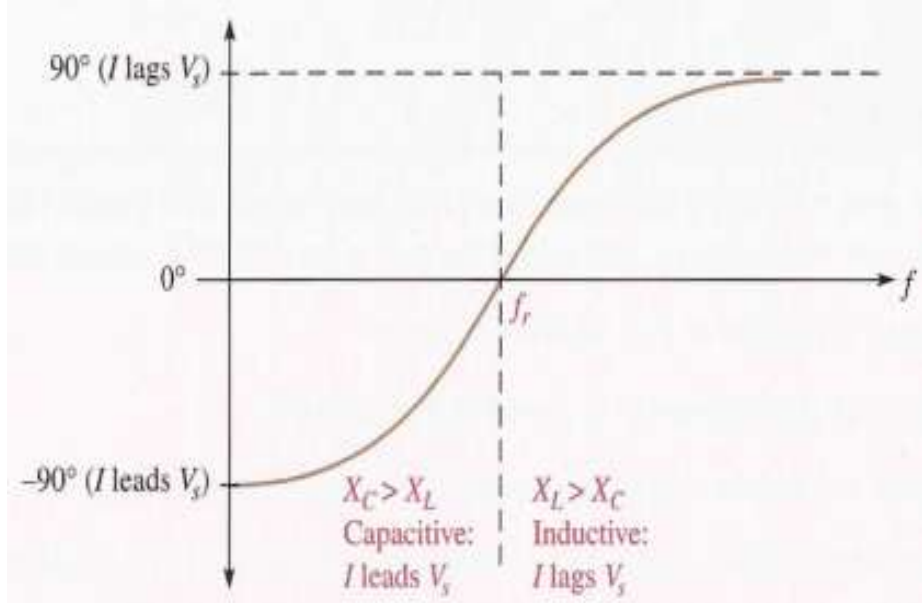
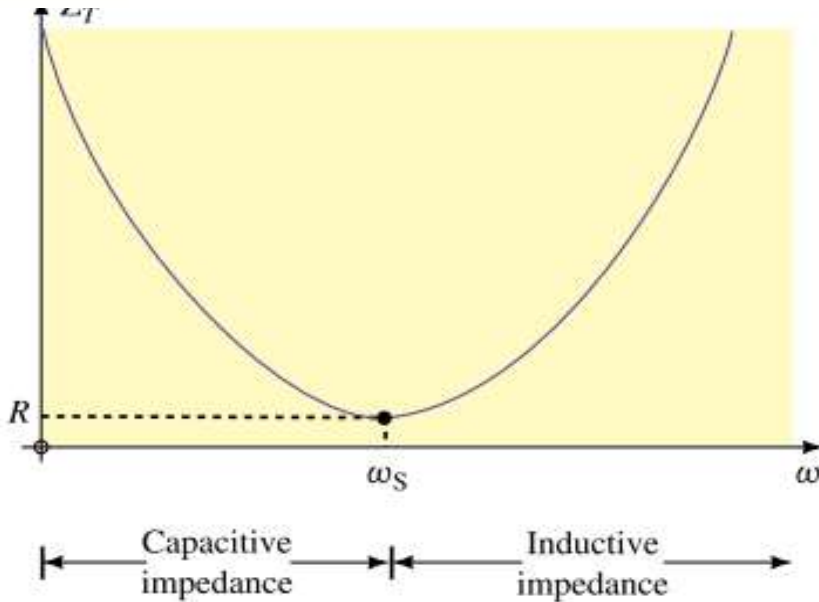
$$Z_T = R$$

$$\theta = \tan^{-1}0 = 0^\circ$$



Impedance of a Series Resonant Circuit **versus** Frequency

FIGURE 21-7 Impedance (magnitude and phase angle) versus angular frequency for a series resonant circuit.



(a) Below f_r , I leads V_s .

(b) At f_r , I is in phase with V_s .

(c) Above f_r , I lags V_s .

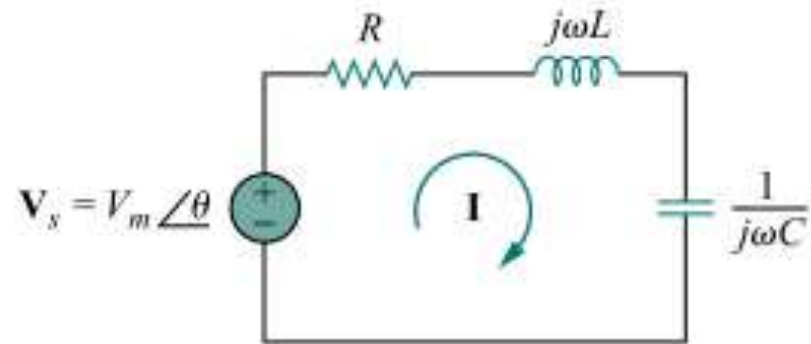
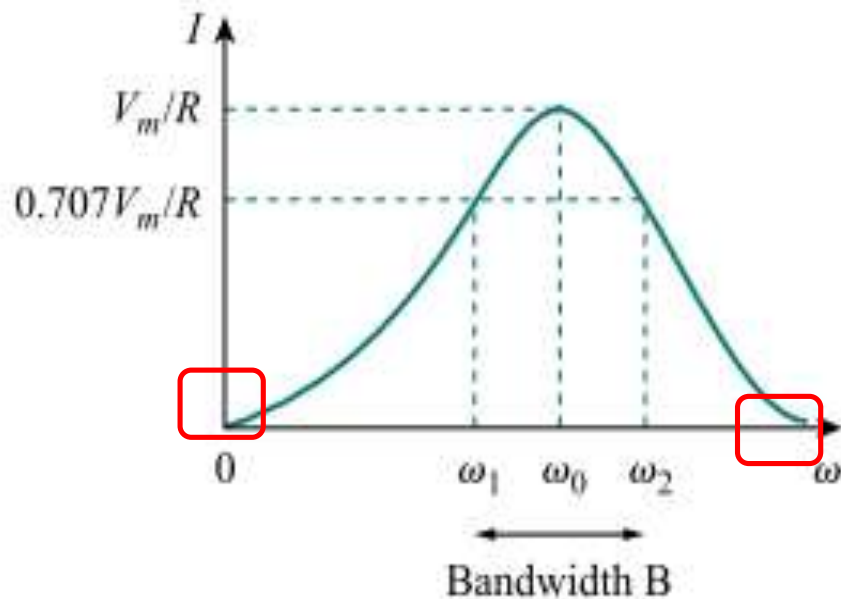
Current and Power in a Series Resonant Circuit

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In this section, we examine how current and power are affected by changing the frequency of the voltage source.

Applying Ohm's law gives the magnitude of the current at resonance as

$$I_{\max} = \frac{V_m}{R}$$



The frequency response of the circuit's current magnitude

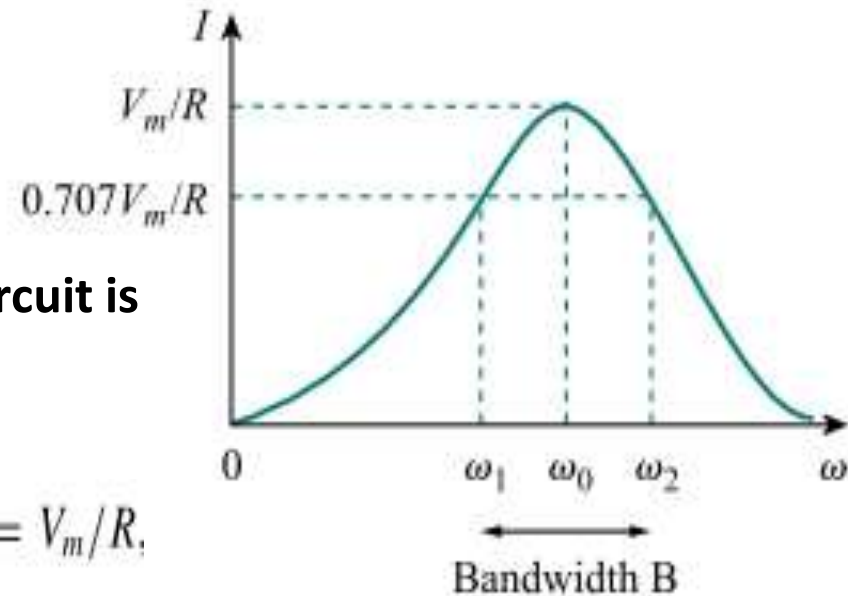
$$I = |\mathbf{I}| = \frac{V_m}{\sqrt{R^2 + (\omega L - 1/\omega C)^2}}$$

For all other frequencies, the magnitude of the current will be less than I_{\max} because the impedance is greater than at resonance.

Current and Power in a Series Resonant Circuit

P.602.C

Since the current is maximum at resonance, it follows that the power must similarly be **maximum** at **resonance**.



➤ The average power dissipated by the RLC circuit is

$$P(\omega) = \frac{1}{2} I^2 R$$

The highest power dissipated occurs at resonance, when $I = V_m/R$,

$$P(\omega_0) = \frac{1}{2} \frac{V_m^2}{R}$$

➤ At certain frequencies $\omega = \omega_1, \omega_2$, the dissipated power is half of that max

$$P(\omega_1) = P(\omega_2) = \frac{V_m^2}{4R}$$

They called the half-power frequencies (Points)

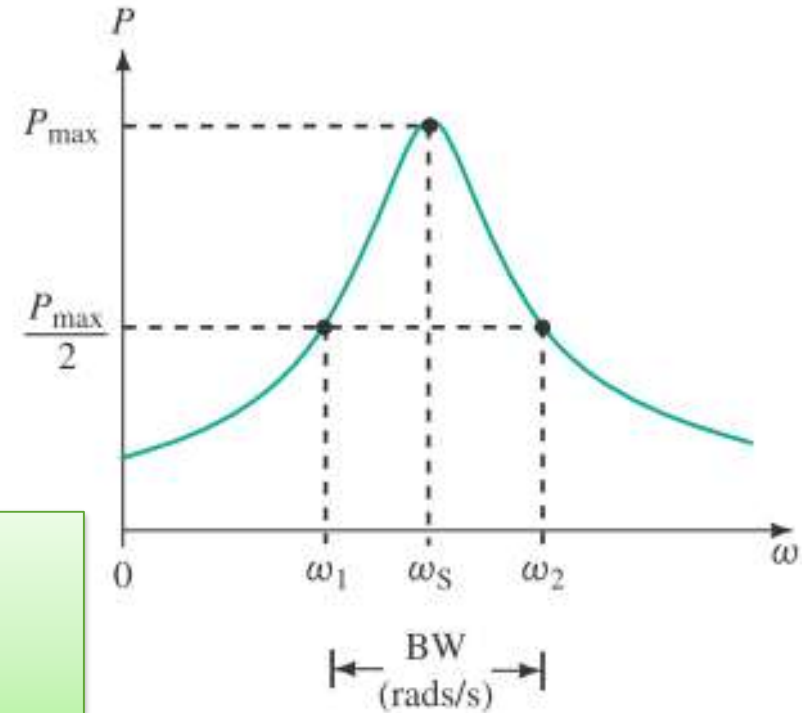


The Bandwidth – Selectivity – Quality Factor

Half-Power Frequencies (Points)
Cutoff Frequencies
Band frequencies

✓ This bell-shaped curve is called the **selectivity curve**

✓ Examining this curve, we see that only frequencies around ω_s will permit significant amounts of power.



The Bandwidth of the resonant circuit (BW)

The difference between the frequencies at which the circuit delivers half of the maximum power.


$$BW = \omega_2 - \omega_1$$

It is called Half-Power Bandwidth

The Bandwidth – Selectivity – Quality Factor

- ✓ If the **bandwidth** of a circuit is kept **very narrow**, the circuit is said to have a **high selectivity**,

since it is highly selective to signals within a very narrow range of frequencies.

- ✓ On the other hand, if the bandwidth of a circuit is **large**, the circuit is said to have a **low selectivity**.

The elements of a series resonant circuit determine:

- The frequency at which the circuit is resonant
- The **shape** (and hence the **bandwidth**) of the power response curve.

1. If R and ω_S are kept constant:

- ✓ By **increasing the ratio of L/C** , the **sides** of the power response curve become **steeper** (i.e. **decrease** in the bandwidth)
- ✓ Inversely, **decreasing** the ratio of L/C causes the sides of the curve to become more gradual (i.e. **increased** bandwidth).

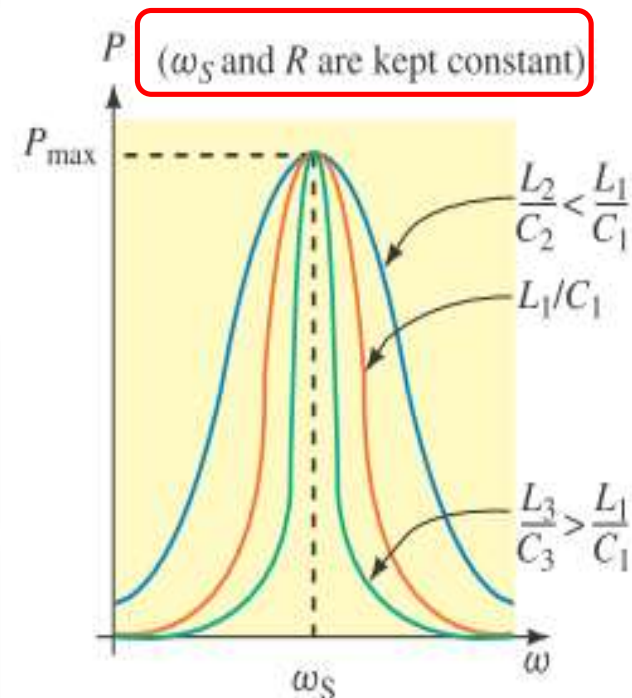
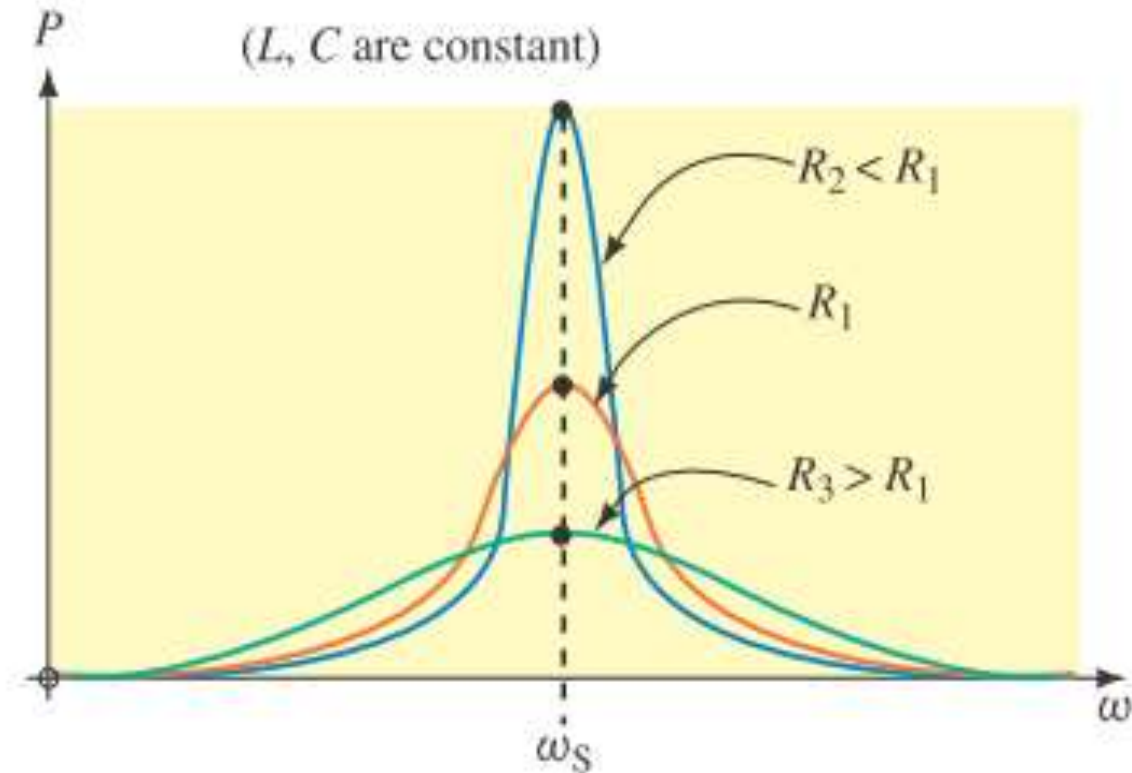


FIGURE 21-10

The Bandwidth – Selectivity – Quality Factor

2. If L and C are kept constant:



- ✓ The **bandwidth** is directly proportional to R
- ✓ The **height** of the curve is **inversely** proportional to R



A series circuit has the **highest selectivity** if the **resistance** of the circuit is kept to a **minimum**.

The Bandwidth – Selectivity – Quality Factor

The half-power frequencies are obtained by setting Z equal to $\sqrt{2}R$,

$$\sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2} = \sqrt{2}R$$

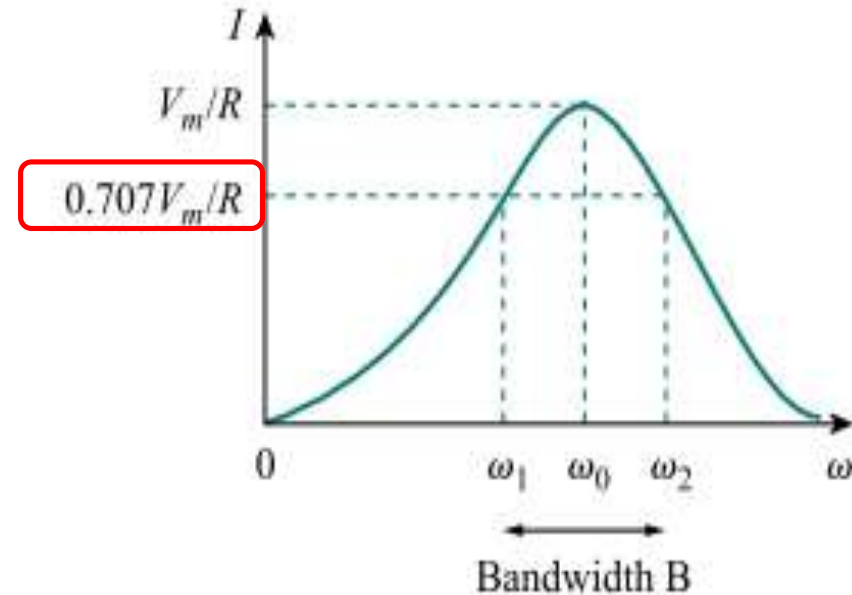
Solving for ω , we obtain

$$\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$
$$\omega_2 = \frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

$$\text{BW} = \omega_2 - \omega_1$$
$$= \frac{R}{2L} + \sqrt{\frac{R^2}{4L^2} + \frac{1}{LC}} - \left(-\frac{R}{2L} + \sqrt{\frac{R^2}{4L^2} + \frac{1}{LC}}\right)$$

$$\text{BW} = \frac{R}{L} \quad (\text{rad/s})$$

$$\omega_0 = \sqrt{\omega_1 \omega_2}$$



The resonant frequency is the **geometric** mean of the half-power frequencies.



The Bandwidth – Selectivity – Quality Factor

- The “sharpness” of the resonance in a resonant circuit is measured quantitatively by the **quality factor Q**.

Q: relates the maximum or peak energy stored to the energy dissipated in the circuit per cycle of oscillation

$$Q = 2\pi \frac{\text{Peak energy stored in the circuit}}{\text{Energy dissipated by the circuit in one period at resonance}}$$

$$Q = \frac{\text{reactive power}}{\text{average power}}$$

Notice that the quality factor is dimensionless.

Q_L is equal to the Q_C at resonance,

$$Q = 2\pi \frac{\frac{1}{2}LI^2}{\frac{1}{2}I^2R(1/f)} = \frac{2\pi fL}{R}$$

$$Q_S = \frac{I^2X_L}{I^2R} = \frac{X_L}{R} = \frac{\omega L}{R}$$

$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 C R}$$



The Bandwidth – Selectivity – Quality Factor

- The relationship between the bandwidth B and the quality factor Q :

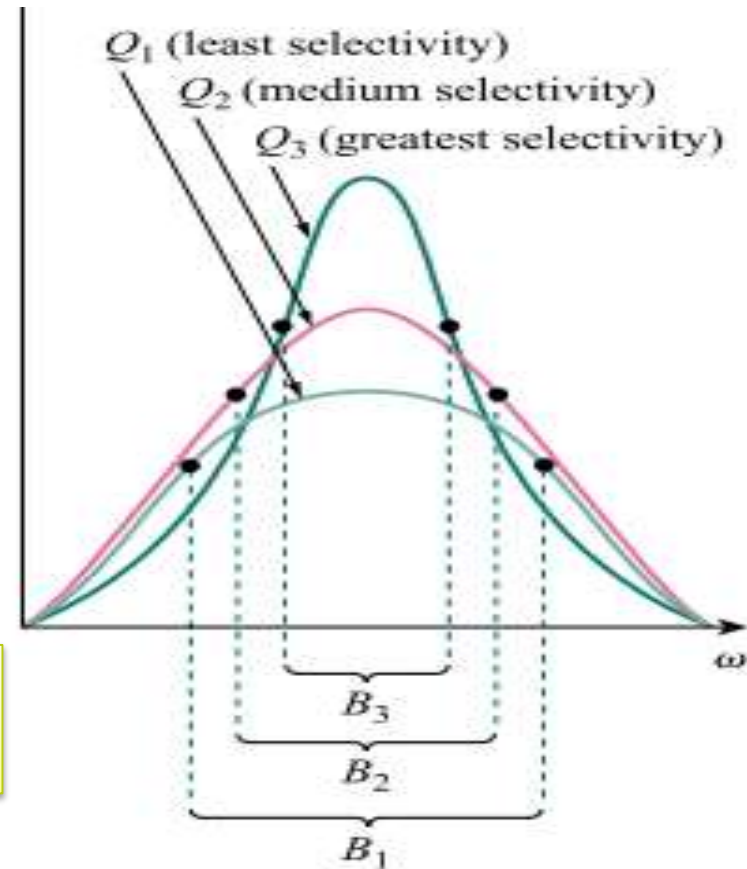
$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 C R}$$

So

$$B = \frac{R}{L} = \frac{\omega_0}{Q} = \omega_0^2 C R$$

$$Q = \frac{\omega_0}{B}$$

The quality factor of a resonant circuit is the ratio of its resonant frequency to its bandwidth.



- The **higher** the value of Q , the **more selective** the circuit is but the **smaller** the bandwidth.

The Bandwidth – Selectivity – Quality Factor

The selectivity of an RLC circuit is the ability of the circuit to respond to a certain frequency and discriminate against all other frequencies.

If the band of frequencies to be selected or rejected is narrow, the **quality factor of the resonant circuit must be high.**

high-Q means equal to or greater than 10.

High-Q circuits are used often in communications networks.

For high-Q, the power frequencies are, for all practical purposes, **symmetrical** around the resonant frequency and can be approximated as:

$$\omega_1 \simeq \omega_0 - \frac{B}{2}, \quad \omega_2 \simeq \omega_0 + \frac{B}{2}$$



Series Resonance Circuit (Cont.)

Quality Factor (Different Formulas)

$$Q_s = \frac{\omega_s L}{R}$$

$$\begin{aligned} Q_s &= \frac{\omega_s L}{R} = \frac{2\pi f_s L}{R} = \frac{2\pi}{R} \left(\frac{1}{2\pi\sqrt{LC}} \right) L \\ &= \frac{L}{R} \left(\frac{1}{\sqrt{LC}} \right) = \left(\frac{\sqrt{L}}{\sqrt{L}} \right) \frac{L}{R\sqrt{LC}} \end{aligned}$$

$$Q_s = \frac{1}{R} \sqrt{\frac{L}{C}}$$



Series Resonance Circuit (Cont.)

EXAMPLE 20.5 A series R - L - C circuit is designed to resonant at $\omega_s = 10^5$ rad/s, have a bandwidth of $0.15\omega_s$, and draw 16 W from a 120-V source at resonance.

- a. Determine the value of R .
- b. Find the bandwidth in hertz.
- c. Find the nameplate values of L and C .
- d. Determine the Q_s of the circuit.

$$BW = f_2 - f_1 = \frac{R}{2\pi L}$$

a. $P = \frac{E^2}{R}$ and $R = \frac{E^2}{P} = \frac{(120 \text{ V})^2}{16 \text{ W}} = \mathbf{900 \ \Omega}$

b. $BW = 0.15f_s$ $f_s = \frac{\omega_s}{2\pi} = \frac{10^5 \text{ rad/s}}{2\pi} = 15,915.49 \text{ Hz}$

$BW = 0.15f_s = 0.15(15,915.49 \text{ Hz}) = \mathbf{2387.32 \text{ Hz}}$

c. $BW = \frac{R}{2\pi L}$ and $L = \frac{R}{2\pi BW} = \frac{900 \ \Omega}{2\pi(2387.32 \text{ Hz})} = \mathbf{60 \text{ mH}}$

$f_s = \frac{1}{2\pi\sqrt{LC}}$ and $C = \frac{1}{4\pi^2 f_s^2 L} = \frac{1}{4\pi^2 (15,915.49 \text{ Hz})^2 (60 \times 10^{-3} \text{ H})}$
 $= \mathbf{1.67 \text{ nF}}$

d. $Q_s = \frac{X_L}{R} = \frac{2\pi f_s L}{R} = \frac{2\pi(15,915.49 \text{ Hz})(60 \text{ mH})}{900 \ \Omega} = \mathbf{6.67}$

Note that at resonance:

1. The impedance is purely resistive, thus, $\mathbf{Z} = R$. In other words, the LC series combination acts like a short circuit, and the entire voltage is across R .
2. The voltage \mathbf{V}_s and the current \mathbf{I} are in phase, so that the power factor is unity.
3. The inductor voltage and capacitor voltage can be much more than the source voltage.

➤ Point (3) can be verified by applying the voltage divider rule to the circuit of Fig. 20.2, we obtain

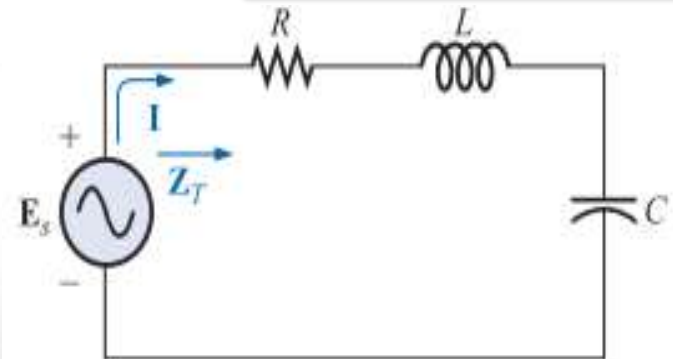
$$V_L = \frac{X_L E}{Z_T} = \frac{X_L E}{R}$$

(at resonance)

$$V_{L_s} = Q_s E$$

$$V_C = \frac{X_C E}{Z_T} = \frac{X_C E}{R}$$

$$V_{C_s} = Q_s E$$



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$$V_{L_s} = Q_s E$$

$$V_{C_s} = Q_s E$$

- Since Q_s is usually greater than 1, the voltage across the capacitor or inductor of a series resonant circuit can be significantly greater than the input voltage.

